

# Analysis of Error Floors of Non-binary LDPC Codes over MBIOS channel

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# Outline

## 1 Background

- Non-binary low-density parity-check (LDPC) codes
- Existing methods for lowering error floors
- Channel model

## 2 Lowering decoding error rates in the error floor over MBIOS channel

- Condition for successful decoding
- Code design method to lower error floor
- Simulation results

## 3 Analysis of symbol error rates in error floors

- Lower bound on symbol error rate in error floor
- Simulation results

## 4 Conclusion and related works

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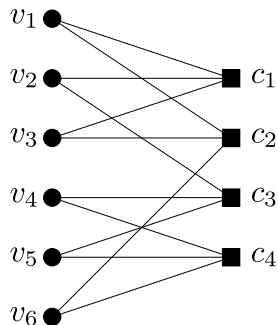
## 4 Conclusion and related works

# Non-binary low-density parity-check (LDPC) codes

## Non-binary LDPC code

Linear code defined by a sparse parity check matrix  $H \in \mathbb{F}_{2^m}^{M \times N}$

$$C := \{\mathbf{x} \in \mathbb{F}_{2^m}^N \mid H\mathbf{x} = \mathbf{0}\}$$



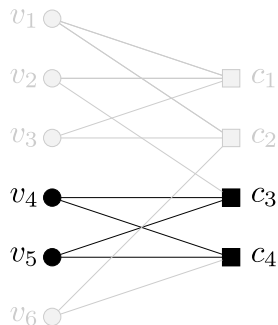
$$\begin{array}{r} \begin{array}{cccccc} & v_1 & v_2 & v_3 & v_4 & v_5 & v_6 \\ c_1 & ( & h_{1,1} & h_{1,2} & h_{1,3} & 0 & 0 & 0 \\ c_2 & & h_{2,1} & 0 & h_{2,3} & 0 & 0 & h_{2,6} \\ c_3 & & 0 & h_{3,2} & 0 & h_{3,4} & h_{3,5} & 0 \\ c_4 & & 0 & 0 & 0 & h_{4,4} & h_{4,5} & h_{4,6} \end{array} \end{array}$$

# Non-binary low-density parity-check (LDPC) codes

## Non-binary LDPC code

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	$v_1$	$v_2$	$v_3$	$v_4$	$v_5$	$v_6$
$c_1$	$h_{1,1}$	$h_{1,2}$	$h_{1,3}$	0	0	0
$c_2$	$h_{2,1}$	0	$h_{2,3}$	0	0	$h_{2,6}$
$c_3$	0	$h_{3,2}$	0	$h_{3,4}$	$h_{3,5}$	0
$c_4$	0	0	0	$h_{4,4}$	$h_{4,5}$	$h_{4,6}$

## Zigzag cycle

Cycle constructed by the variable nodes of degree 2

## Zigzag cycles degrade decoding performance

- Error floors are mainly caused by codewords of small weight.

### Example : Zigzag cycle yields codeword

$\alpha$  : primitive element of  $\mathbb{F}_{2^m}$

$$H = \begin{pmatrix} h_{1,1} & h_{1,2} & h_{1,3} & 0 & 0 & 0 \\ h_{2,1} & 0 & h_{2,3} & 0 & 0 & h_{2,6} \\ 0 & h_{3,2} & 0 & 1 & \alpha^4 & 0 \\ 0 & 0 & 0 & \alpha^6 & \alpha^{10} & h_{4,6} \end{pmatrix}$$

$$(0 \ 0 \ 0 \ \alpha^4 \ 1 \ 0) \in C$$

$$\det \begin{pmatrix} 1 & \alpha^4 \\ \alpha^6 & \alpha^{10} \end{pmatrix} = 0$$

Zigzag cycle yields codeword.

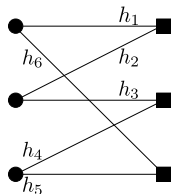
# Cycle cancellation [Poulliat 2008]

## Existing method : Cycle cancellation [Poulliat 2008]

Design edge labels in zigzag cycle so that

$$\det \tilde{H} \neq 0.$$

$$\tilde{H} = \begin{pmatrix} h_1 & h_2 & 0 \\ 0 & h_3 & h_4 \\ h_6 & 0 & h_5 \end{pmatrix}$$



$$\det \tilde{H} = h_1 h_3 h_5 + h_2 h_4 h_6 \neq 0 \iff \beta = h_1^{-1} h_2 h_3^{-1} h_4 h_5^{-1} h_6 \neq 1$$

$$\iff \beta = \prod_{i=1}^3 h_{2i-1}^{-1} h_{2i} \neq 1$$

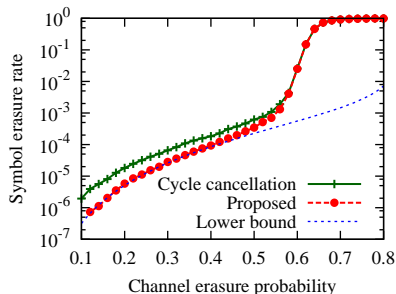
[Poulliat 2008] C. Poulliat, M. Fossorier and D. Declercq, "Design of regular  $(2, d_c)$ -LDPC codes over  $\text{GF}(q)$  using their binary images", IEEE Trans. Comm. (2008)

## Binary erasure channel (BEC) case [ISIT 2010]

Existing method: Modified cycle cancellation [ISIT 2010]

Design edge labels in zigzag cycle so that

$$\beta \notin \bigcup_{r>0:r|m, r \neq m} \left\{ \alpha^{i \frac{2^m-1}{2^r-1}} \mid i = 0, 1, \dots, 2^r - 2 \right\}$$



Symbol code length : 315

Order of Galois field : 16

[ISIT 2010] T. Nozaki, K. Kasai and K. Sakaniwa, "Error floors of non-binary LDPC codes," ISIT 2010



# Channel Model

## Memoryless binary-input output-symmetric (MBIOS) channel

- Memoryless

$$p_{Y|X}(y | x) = \prod_t p_{Y_t|X_t}(y_t | x_t)$$

- symmetric

$$p_{Y_t|X_t}(y_t | 0) = p_{Y_t|X_t}(-y_t | 1)$$

## Example : Member of MBIOS channel

- Binary erasure channel (BEC)
- Binary symmetric channel (BSC)
- Additive white Gaussian noise (AWGN) channel

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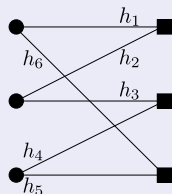
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# Condition for successful decoding

## Definition : zigzag cycle code



symbol code length 3

$$\beta = \prod_{i=1}^s h_{2i-1}^{-1} h_{2i}$$

$$\gamma_i = h_{2i-1}^{-1} h_{2i}$$

$s$  : symbol code length

$\sigma$  : the order of  $\beta$

$C_k(\gamma)$  :  $\gamma$ -th entry of the initial message of  $k$ -th variable node

## Lemma 1 : Condition for successful decoding

BP decoding succeeds for zigzag cycle code of length  $s$

$\iff$  for all  $x \in A_\beta := \{\alpha^j \mid j = 0, 1, \dots, \frac{2^m-1}{\sigma} - 1\}$

$$\prod_{k=1}^s (C_k(0))^\sigma > \prod_{t=0}^{\sigma-1} \prod_{k=1}^s C_k \left( \beta^t x \prod_{s=1}^{k-1} \gamma_s \right).$$

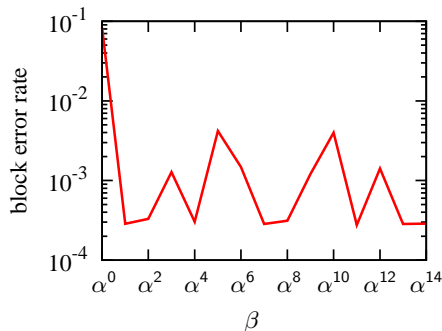
# Condition for the lowest decoding error rate

## Theorem 1

The zigzag cycles with cycle parameter  $\beta \notin \mathcal{H}_m$  have good decoding performance, where

$$\mathcal{H}_m := \bigcup_{0 < r < 2^m - 1 : r | 2^m - 1} \left\{ \alpha^{i \frac{2^m - 1}{r}} \mid i = 0, \dots, r - 1 \right\}.$$

## Simulation result



additive white Gaussian noise  
(AWGN) channel

$$\sigma^2 = 1$$

Zigzag cycle code

Symbol code length : 3

Order of Galois field : 16

## Construction of code

**Proposition 1: Design method to lower error floor for MBIOS channel**

Design edge labels in zigzag cycle so that

$$\beta \notin \mathcal{H}_m = \bigcup_{0 < r < 2^m - 1: r | 2^m - 1} \left\{ \alpha^{i \frac{2^m - 1}{r}} \mid i = 0, \dots, r - 1 \right\}.$$

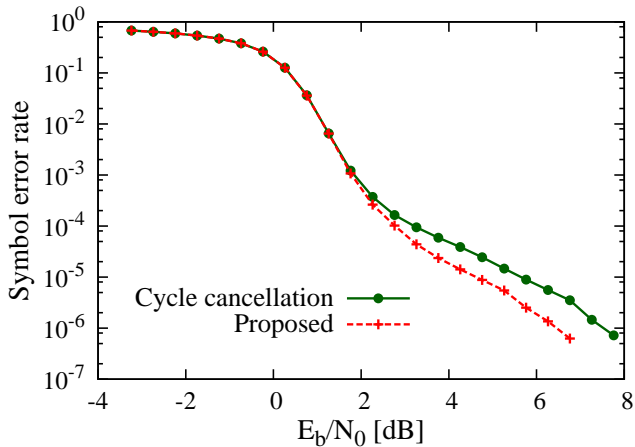
**Note : BEC case**

Design edge labels in zigzag cycle so that

$$\beta \notin \bigcup_{r > 0: r | m, r \neq m} \left\{ \alpha^{i \frac{2^m - 1}{2^r - 1}} \mid i = 0, \dots, r - 1 \right\}.$$

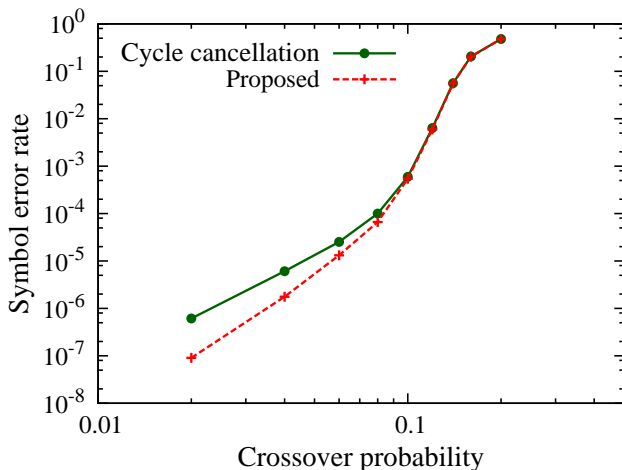
Field	The elements of $\mathcal{H}_m$
$\mathbb{F}_{2^4}$	$1, \alpha^3, \alpha^5, \alpha^6, \alpha^9, \alpha^{10}, \alpha^{12}$
$\mathbb{F}_{2^6}$	$1, \alpha^3, \alpha^6, \alpha^7, \alpha^9, \alpha^{12}, \alpha^{14}, \alpha^{15}, \alpha^{18}, \alpha^{21}, \alpha^{24}, \alpha^{25}, \alpha^{27}, \alpha^{28}, \alpha^{30}, \alpha^{33}, \alpha^{35}, \alpha^{36}, \alpha^{39}, \alpha^{42}, \alpha^{45}, \alpha^{48}, \alpha^{49}, \alpha^{51}, \alpha^{54}, \alpha^{56}, \alpha^{57}, \alpha^{60}$

## Simulation result (AWGN channel)



(2,3)-regular LDPC code ensemble  
Symbol code length : 315 symbols  
Order of Galois field : 16

## Simulation result (BSC)



(2,3)-regular LDPC code ensemble  
Symbol code length : 315 symbols  
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## Condition for successful decoding for zigzag cycle codes

### Definition : Log-likelihood ratio

$Y_{v,i}$  :  $i$ -th channel output in  $v$ -th variable node

$$Z_{v,i} := \log \frac{\Pr(Y_{v,i} | X_{v,i} = 0)}{\Pr(Y_{v,i} | X_{v,i} = 1)}.$$

### All zero assumption

The decoding error rate does not depend on the codeword.

### Lemma2 : Condition for successful decoding

For zigzag cycle codes with cycle parameter  $\beta \notin \mathcal{H}_m$  under BP decoding

All the symbols are successfully decoded  $\iff \sum_{v,i} Z_{v,i} > 0$ .

All the errors are not recoverable  $\iff \sum_{v,i} Z_{v,i} \leq 0$ .

The decoding error probabilities of all the zigzag cycle codes are lower bounded by  $\Pr(\sum_{v,i} Z_{v,i} \geq 0)$

## Analysis of error floors

### Theorem 2 : Error floor for NB-LDPC codes

$N$  : symbol code length

$\lambda(x) = \sum_i \lambda_i x^{i-1}$ ,  $\rho(x) := \sum_i \rho_i x^{i-1}$  : degree distribution pair

$\mu := \lambda'(0)\rho'(1)$

$s_g$  : expurgation parameter (no cycle of weight in  $\{1, 2, \dots, s_g - 1\}$ )

$Z^{(sm)} := \sum_{v=1}^s \sum_{i=1}^m Z_{v,i}$

Let  $P_s(\mathcal{E})$  be the symbol error rate of the code ensemble designed by proposed method over the MBIOS channel characterized by its  $L$ -density  $\mathbf{a}$  under BP decoding.

For  $\int \mathbf{a}(x) \exp[-x/2] dx < \mu^{-\frac{1}{m}}$  and the sufficiently large  $N$ , we have

$$P_s(\mathcal{E}) \geq \frac{1}{2N} \sum_{s=s_g}^{\infty} \mu^s \Pr(Z^{(sm)} \leq 0) + o\left(\frac{1}{N}\right).$$

## Analysis of error floors

### Corollary1 : AWGN channel case

For AWGN channel with channel variance  $\sigma^2$

$$P_s(\mathcal{E}) \geq \frac{1}{2N} \sum_{s=s_g}^{\infty} \mu^s Q\left(\frac{\sqrt{sm}}{\sigma}\right) + o\left(\frac{1}{N}\right), \quad \text{for } \sigma^2 < \begin{cases} \infty & \text{for } \mu \leq 1 \\ \frac{m}{2 \log \mu} & \text{for } \mu > 1 \end{cases},$$

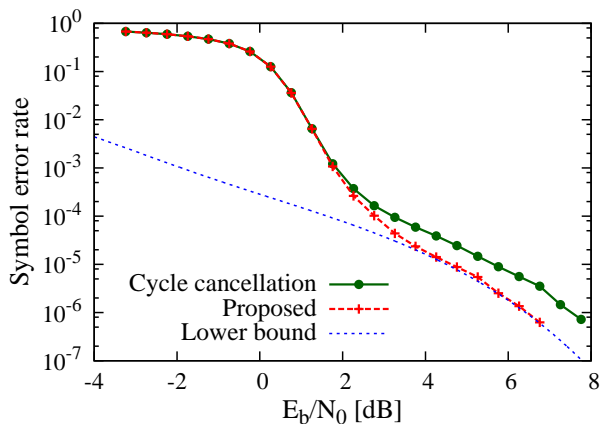
where  $Q(y) = \frac{1}{\sqrt{2\pi}} \int_y^{\infty} \exp[-\frac{x^2}{2}] dx$ .

### Corollary2 : BSC case

For BSC with crossover probability  $\epsilon < \frac{1 - \sqrt{1 - \max\{\mu^{-2/m}, 1\}}}{2}$ ,

$$P_s(\mathcal{E}) \geq \frac{1}{2N} \sum_{s=s_c}^{\infty} \mu^s \sum_{i \leq ms/2} \binom{ms}{i} (1 - \epsilon)^{ms-i} \epsilon^i + o\left(\frac{1}{N}\right).$$

## Simulation result : Lower bound (AWGN channel)



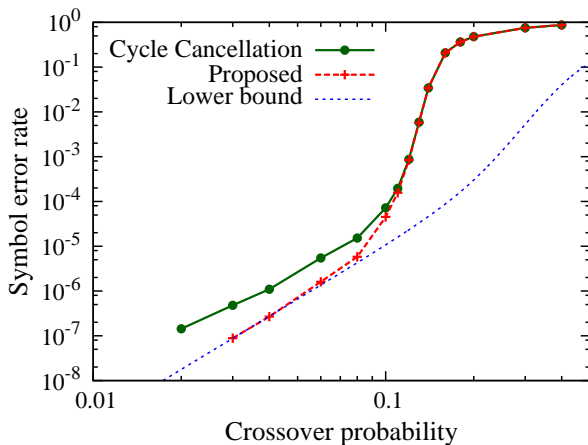
(2,3)-regular LDPC code ensemble

Symbol code length : 315

Order of Galois field : 16

$$s_g = 1$$

## Simulation result : Lower bound (BSC)



(2,3)-regular LDPC code ensemble

Symbol code length : 1200

Order of Galois field : 16

$$s_g = 2$$

# Conclusion and related works

## Conclusion

- We derive a necessary and sufficient condition for successful decoding for zigzag cycle codes over the MBIOS channel under BP decoding.
- We propose a design method lowering error floor.
- We give a lower bound for the decoding error rates in the error floors for the expurgated ensembles constructed by proposed method.

## Related works

- Analysis of error floors of non-binary LDPC codes over  $q$ -ary symmetric channel
- Analysis of error floors of non-binary LDPC codes defined over general linear groups